

# LMK02000

## Precision Clock Conditioner with Integrated PLL

### General Description

The LMK02000 precision clock conditioner combines the functions of jitter cleaning/reconditioning, multiplication, and distribution of a reference clock. The device integrates a high performance Integer-N Phase Locked Loop (PLL), three LVDS, and five LVPECL clock output distribution blocks.

Each clock distribution block includes a programmable divider, a phase synchronization circuit, a programmable delay, a clock output mux, and an LVDS or LVPECL output buffer. This allows multiple integer-related and phase-adjusted copies of the reference to be distributed to eight system components.

The clock conditioner comes in a 48-pin LLP package and is footprint compatible with other clocking devices in the same family.

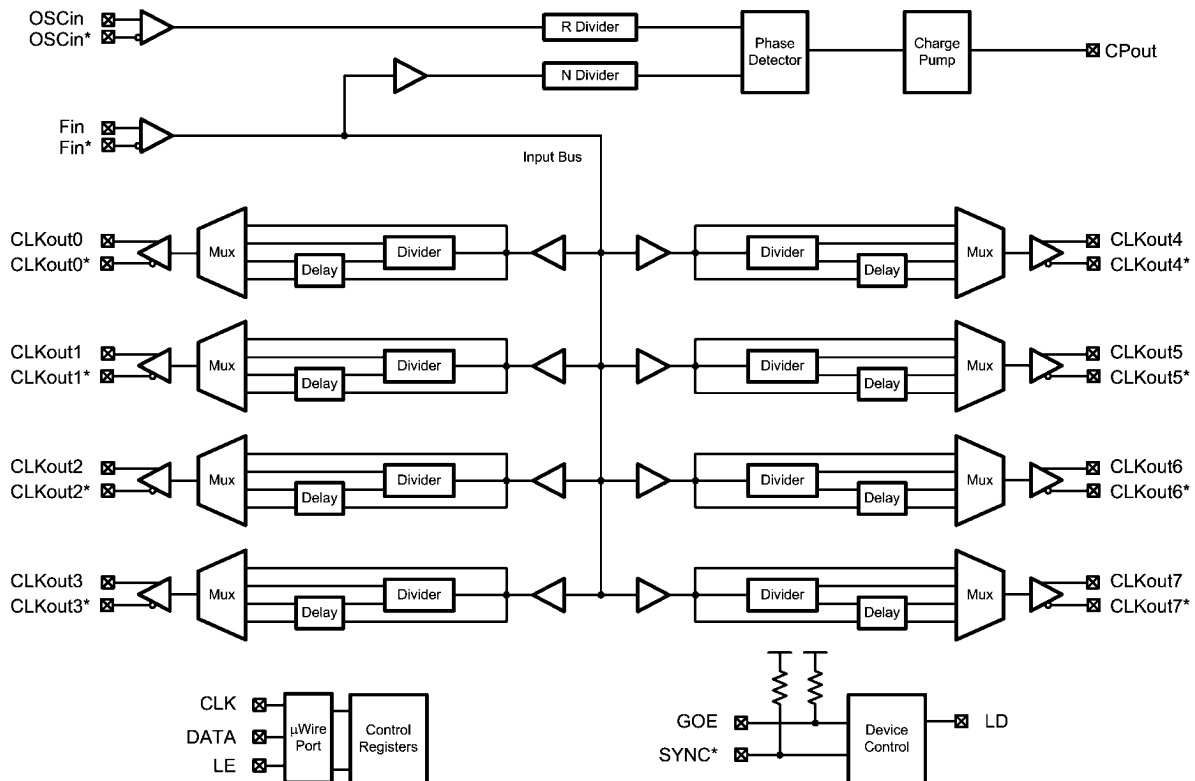
### Features

- 40 fs additive jitter
- Integrated Integer-N PLL
- Clock output frequency range of 1 to 800 MHz
- 3 LVDS and 5 LVPECL clock outputs
- Dedicated divider and delay blocks on each clock output
- Pin compatible family of clocking devices
- 3.15 to 3.45 V operation
- Package: 48 pin LLP (7.0 x 7.0 x 0.8 mm)

### Target Applications

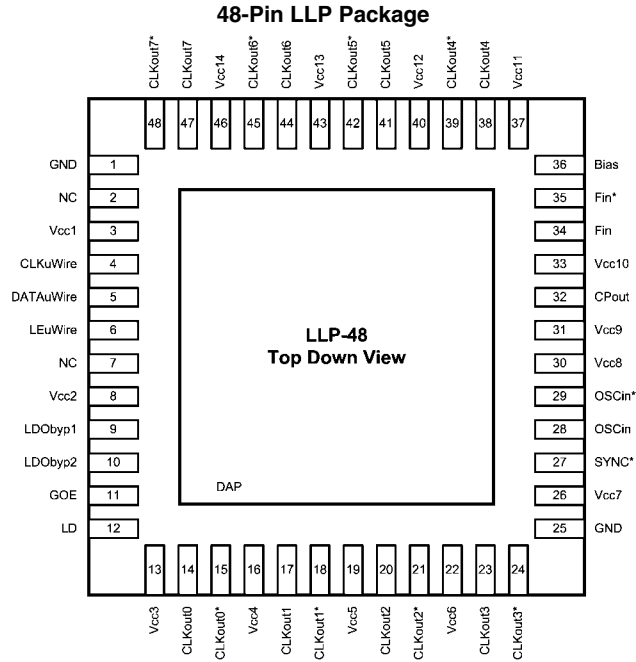
- Data Converter Clocking
- SONET/SDH, DSLAM
- Networking
- Wireless Infrastructure
- Medical
- Test and Measurement
- Military / Aerospace

### Functional Block Diagram



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# Connection Diagram



20216502

## Pin Descriptions

Pin #	Pin Name	I/O	Description
1, 25	GND	-	Ground
2, 7	NC	-	No Connection to these pins
3, 8, 13, 16, 19, 22, 26, 30, 31, 33, 37, 40, 43, 46	Vcc1, Vcc2, Vcc3, Vcc4, Vcc5, Vcc6, Vcc7, Vcc8, Vcc9, Vcc10, Vcc11, Vcc12, Vcc13, Vcc14	-	Power Supply
4	CLKuWire	I	MICROWIRE Clock Input
5	DATAuWire	I	MICROWIRE Data Input
6	LEuWire	I	MICROWIRE Latch Enable Input
9, 10	LDObyp1, LDObyp2	-	LDO Bypass
11	GOE	I	Global Output Enable
12	LD	O	Lock Detect and Test Output
14, 15	CLKout0, CLKout0*	O	LVDS Clock Output 0
17, 18	CLKout1, CLKout1*	O	LVDS Clock Output 1
20, 21	CLKout2, CLKout2*	O	LVDS Clock Output 2
23, 24	CLKout3, CLKout3*	O	LVPECL Clock Output 3
27	SYNC*	I	Global Clock Output Synchronization
28, 29	OSCin, OSCin*	I	Oscillator Clock Input; Must be AC coupled
32	CPout	O	Charge Pump Output
34, 35	Fin, Fin*	I	Frequency Input; Must be AC coupled
36	Bias	I	Bias Bypass
38, 39	CLKout4, CLKout4*	O	LVPECL Clock Output 4
41, 42	CLKout5, CLKout5*	O	LVPECL Clock Output 5
44, 45	CLKout6, CLKout6*	O	LVPECL Clock Output 6
47, 48	CLKout7, CLKout7*	O	LVPECL Clock Output 7
DAP	DAP	-	Die Attach Pad is Ground

## Absolute Maximum Ratings (Notes 1, 2)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Parameter	Symbol	Ratings	Units
Power Supply Voltage	$V_{CC}$	-0.3 to 3.6	V
Input Voltage	$V_{IN}$	-0.3 to ( $V_{CC} + 0.3$ )	V
Storage Temperature Range	$T_{STG}$	-65 to 150	°C
Lead Temperature (solder 4 s)	$T_L$	+260	°C

## Recommended Operating Conditions

Parameter	Symbol	Min	Typ	Max	Units
Ambient Temperature	$T_A$	-40	25	85	°C
Power Supply Voltage	$V_{CC}$	3.15	3.3	3.45	V

**Note 1:** Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Recommended Operating Conditions indicate conditions for which the device is intended to be functional, but do not guarantee specific performance limits. For guaranteed specifications and test conditions, see the Electrical Characteristics. The guaranteed specifications apply only to the test conditions listed.

**Note 2:** This device is a high performance integrated circuit with ESD handling precautions. Handling of this device should only be done at ESD protected work stations. The device is rated to a HBM-ESD of >2 kV, a MM-ESD of >200 V, and a CDM-ESD of >1.2 kV.

## Electrical Characteristics

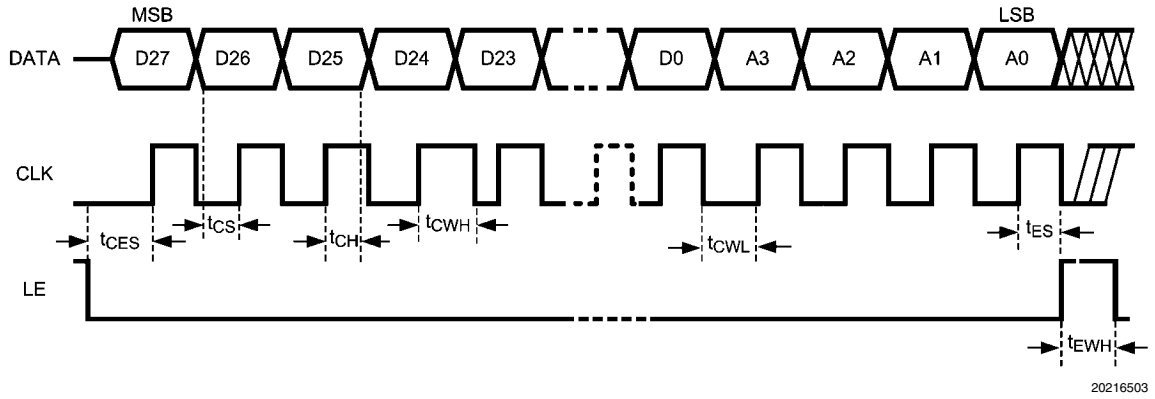
( $V_{CC} = 3.3$  V,  $-40$  °C  $\leq T_A \leq 85$  °C; Differential Inputs/Outputs; except as specified.)

Symbol	Parameter	Conditions	Min	Typ	Max	Units
<b>Current Consumption</b>						
$I_{CC}$	Power Supply Current	Entire device with one LVPECL and one LVDS output operating at 622.08 MHz in bypass mode		155		mA
		Divider Circuitry (CLKoutX_DIV = 4) Added for each clock output		9		
		Delay Circuitry (CLKoutX_DLY = 2250 ps) Added for each clock output		10		
$I_{CC}^{PD}$	Power Down Current	POWERDOWN = 1		1		mA
<b>Reference Oscillator</b>						
$f_{OSCin}$	Reference Oscillator Input Frequency Range		1		200	MHz
$SLEW_{OSCin}$	Reference Oscillator Input Slew Rate	(Note 3)	0.5			V/ns
$V_{OSCin}$	Input voltage for OSCin and OSCin*	AC coupled; Single ended	0.2		1.6	V <sub>pp</sub>
<b>Frequency Input</b>						
$f_{Fin}$	Frequency Input Frequency Range				800	MHz
$SLEW_{Fin}$	Frequency Input Slew Rate	(Note 4)	0.5			V/ns
$DUTY_{Fin}$	Fin Duty Cycle		40		60	%
$P_{Fin}$	Input Power Range for Fin or Fin*	AC coupled; Single ended	-15		8	dBm

Symbol	Parameter	Conditions	Min	Typ	Max	Units	
<b>PLL</b>							
$f_{COMP}$	Phase Detector Frequency				40	MHz	
$I_{SRCECPout}$	Charge Pump Source Current	$V_{CPout} = V_{CC}/2$ , PLL_CP_GAIN = 1x		100		$\mu A$	
		$V_{CPout} = V_{CC}/2$ , PLL_CP_GAIN = 4x		400			
		$V_{CPout} = V_{CC}/2$ , PLL_CP_GAIN = 16x		1600			
		$V_{CPout} = V_{CC}/2$ , PLL_CP_GAIN = 32x		3200			
$I_{SINKCPout}$	Charge Pump Sink Current	$V_{CPout} = V_{CC}/2$ , PLL_CP_GAIN = 1x		-100		$\mu A$	
		$V_{CPout} = V_{CC}/2$ , PLL_CP_GAIN = 4x		-400			
		$V_{CPout} = V_{CC}/2$ , PLL_CP_GAIN = 16x		-1600			
		$V_{CPout} = V_{CC}/2$ , PLL_CP_GAIN = 32x		-3200			
$I_{CPoutTRI}$	Charge Pump TRI-STATE Current	$0.5 V < V_{CPout} < V_{CC} - 0.5 V$		2	10	nA	
$I_{CPout\%MIS}$	Magnitude of Charge Pump Sink vs. Source Current Mismatch	$V_{CPout} = V_{CC} / 2$ $T_A = 25^\circ C$		3		%	
$I_{CPoutVTUNE}$	Magnitude of Charge Pump Current vs. Charge Pump Voltage Variation	$0.5 V < V_{CPout} < V_{CC} - 0.5 V$ $T_A = 25^\circ C$		4		%	
$I_{CPoutTEMP}$	Magnitude of Charge Pump Current vs. Temperature Variation			4		%	
PN10kHz	PLL 1/f Noise at 10 kHz Offset (Note 5) Normalized to 1 GHz Output Frequency	PLL_CP_GAIN = 1x		-117		dBc/Hz	
		PLL_CP_GAIN = 32x		-122			
PN1Hz	Normalized Phase Noise Contribution (Note 6)	PLL_CP_GAIN = 1x		-219		dBc/Hz	
		PLL_CP_GAIN = 32x		-224			
<b>Clock Distribution Section (Note 7) - LVDS Clock Outputs (CLKout0 to CLKout2)</b>							
$Jitter_{ADD}$	Additive RMS Jitter (Note 7)	$R_L = 100 \Omega$ Input Bus = 800 MHz Bandwidth = 12 kHz to 20 MHz	CLKoutX_MUX = Bypass		40		fs
			CLKoutX_MUX = Divided CLKoutX_DIV = 4		150		
$t_{SKEW}$	CLKoutX to CLKoutY	Equal loading and identical channel configuration $R_L = 100 \Omega$	-30	$\pm 4$	30	ps	
$V_{OD}$	Differential Output Voltage	$R_L = 100 \Omega$	250	350	450	mV	
$\Delta V_{OD}$	Change in magnitude of $V_{OD}$ for complementary output states	$R_L = 100 \Omega$	-35		35	mV	
$V_{OS}$	Output Offset Voltage	$R_L = 100 \Omega$	1.125	1.25	1.375	V	
$\Delta V_{OS}$	Change in magnitude of $V_{OS}$ for complementary output states	$R_L = 100 \Omega$	-35		35	mV	
$I_{SA}$ $I_{SB}$	Clock Output Short Circuit Current single ended	Single ended outputs shorted to GND	-24		24	mA	
$I_{SAB}$	Clock Output Short Circuit Current differential	Complementary outputs tied together	-12		12	mA	

Symbol	Parameter	Conditions	Min	Typ	Max	Units
<b>Clock Distribution Section (Note 7) - LVPECL Clock Outputs (CLKout3 to CLKout7)</b>						
$Jitter_{ADD}$	Additive RMS Jitter (Note 7)	$R_L = 100 \Omega$ Input Bus = 800 MHz Integration Bandwidth = 12 kHz to 20 MHz	CLKoutX_MUX = Bypass		40	fs
			CLKoutX_MUX = Divided CLKoutX_DIV = 4		150	
$t_{SKEW}$	CLKoutX to CLKoutY	Equal loading and identical channel configuration Termination = 50 $\Omega$ to Vcc - 2 V	-30	$\pm 3$	30	ps
$V_{OH}$	Output High Voltage	Termination = 50 $\Omega$ to Vcc - 2 V		Vcc - 0.98		V
$V_{OL}$	Output Low Voltage			Vcc - 1.8		V
$V_{OD}$	Differential Output Voltage		660	810	965	mV
<b>Digital LVTTTL Interfaces (Note 8)</b>						
$V_{IH}$	High-Level Input Voltage		2.0		Vcc	V
$V_{IL}$	Low-Level Input Voltage				0.8	V
$I_{IH}$	High-Level Input Current	$V_{IH} = V_{cc}$	-1.0		1.0	$\mu A$
$I_{IL}$	Low-Level Input Current	$V_{IL} = 0$	-1.0		1.0	$\mu A$
$V_{OH}$	High-Level Output Voltage	$I_{OH} = -500 \mu A$		Vcc - 0.4		V
$V_{OL}$	Low-Level Output Voltage	$I_{OL} = -500 \mu A$			0.4	V
<b>Digital MICROWIRE Interfaces (Note 9)</b>						
$V_{IH}$	High-Level Input Voltage		1.6		Vcc	V
$V_{IL}$	Low-Level Input Voltage				0.4	V
$I_{IH}$	High-Level Input Current	$V_{IH} = V_{cc}$	-1.0		1.0	$\mu A$
$I_{IL}$	Low-Level Input Current	$V_{IL} = 0$	-1.0		1.0	$\mu A$
<b>MICROWIRE Timing</b>						
$t_{CS}$	Data to Clock Set Up Time	See Data Input Timing	25			ns
$t_{CH}$	Data to Clock Hold Time	See Data Input Timing	8			ns
$t_{CWH}$	Clock Pulse Width High	See Data Input Timing	25			ns
$t_{CWL}$	Clock Pulse Width Low	See Data Input Timing	25			ns
$t_{ES}$	Clock to Enable Set Up Time	See Data Input Timing	25			ns
$t_{CES}$	Enable to Clock Set Up Time	See Data Input Timing	25			ns
$t_{EWH}$	Enable Pulse Width High	See Data Input Timing	25			ns
<p><b>Note 3:</b> For all frequencies the slew rate, <math>SLEW_{OSCin}</math>, is measured between 20% and 80%. If only OSCin is being driven (OSCin* AC grounded), the slew rate is half, 0.25 V/ns.</p> <p><b>Note 4:</b> For all frequencies the slew rate, <math>SLEW_{Fin}</math>, is measured between 20% and 80%. If only Fin is being driven (Fin* AC grounded), the slew rate is half, 0.25 V/ns.</p> <p><b>Note 5:</b> A specification in modeling PLL in-band phase noise is the 1/f flicker noise, <math>L_{PLL\_flicker}(f)</math>, which is dominant close to the carrier. Flicker noise has a 10 dB/decade slope. <math>PN_{10kHz}</math> is normalized to a 10 kHz offset and a 1 GHz carrier frequency. <math>PN_{10kHz} = L_{PLL\_flicker}(10\text{ kHz}) - 20\log(F_{out} / 1\text{ GHz})</math>, where <math>L_{PLL\_flicker}(f)</math> is the single side band phase noise of only the flicker noise's contribution to total noise, L(f). To measure <math>L_{PLL\_flicker}(f)</math> it is important to be on the 10 dB/decade slope close to the carrier. A high compare frequency and a clean crystal are important to isolating this noise source from the total phase noise, L(f).</p> <p><b>Note 6:</b> A specification in modeling PLL in-band phase noise is the Normalized Phase Noise Contribution, <math>L_{PLL\_flat}(f)</math>, of the PLL and is defined as: <math>PN_{1Hz} = L_{PLL\_flat}(f) - 20\log(N) - 10\log(F_{comp})</math>. <math>L_{PLL\_flat}(f)</math> is the single side band phase noise measured at an offset frequency, f, in a 1 Hz Bandwidth and Fcomp is the comparison frequency of the synthesizer. <math>L_{PLL\_flat}(f)</math> contributes to the total noise, L(f). To measure <math>L_{PLL\_flat}(f)</math> the offset frequency, f, must be chosen sufficiently smaller than the loop bandwidth of the PLL, and yet large enough to avoid a substantial noise contribution from the reference and flicker noise.</p> <p><b>Note 7:</b> The Clock Distribution Section includes all parts of the device except the PLL section. Typical Additive Jitter specifications apply to the clock distribution section only.</p> <p><b>Note 8:</b> Applies to GOE, LD, and SYNC*.</p> <p><b>Note 9:</b> Applies to uWireCLK, uWireDATA, and uWireLE.</p>						

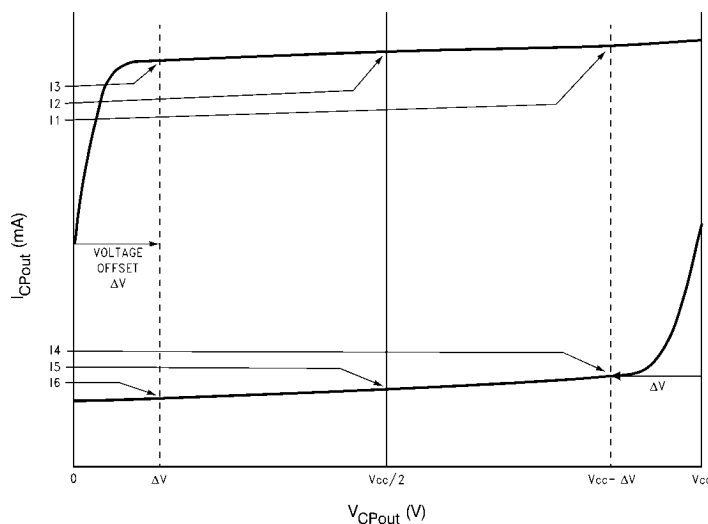
## Serial Data Timing Diagram



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Data bits set on the DATA signal are clocked into a shift register, MSB first, on each rising edge of the CLK signal. On the rising edge of the LE signal, the data is sent from the shift register to the addressed register determined by the LSB bits. After the programming is complete the CLK, DATA, and LE signals should be returned to a low state.

## Charge Pump Current Specification Definitions



20216531

I1 = Charge Pump Sink Current at  $V_{CPout} = V_{CC} - \Delta V$

I2 = Charge Pump Sink Current at  $V_{CPout} = V_{CC}/2$

I3 = Charge Pump Sink Current at  $V_{CPout} = \Delta V$

I4 = Charge Pump Source Current at  $V_{CPout} = V_{CC} - \Delta V$

I5 = Charge Pump Source Current at  $V_{CPout} = V_{CC}/2$

I6 = Charge Pump Source Current at  $V_{CPout} = \Delta V$

$\Delta V$  = Voltage offset from the positive and negative supply rails. Defined to be 0.5 V for this device.

### Charge Pump Output Current Magnitude Variation vs. Charge Pump Output Voltage

$$I_{CPout} \text{ Vs } V_{CPout} = \frac{|I1| - |I3|}{|I1| + |I3|} \times 100\%$$

$$= \frac{|I4| - |I6|}{|I4| + |I6|} \times 100\%$$

20216532

### Charge Pump Sink Current vs. Charge Pump Output Source Current Mismatch

$$I_{CPout} \text{ Sink Vs } I_{CPout} \text{ Source} = \frac{|I2| - |I5|}{|I2| + |I5|} \times 100\%$$

20216533

### Charge Pump Output Current Magnitude Variation vs. Temperature

$$I_{CPout} \text{ Vs } T_A = \frac{|I2|_{T_A=25^\circ C} - |I2|_{T_A=25^\circ C}}{|I2|_{T_A=25^\circ C}} \times 100\%$$

$$= \frac{|I5|_{T_A=25^\circ C} - |I5|_{T_A=25^\circ C}}{|I5|_{T_A=25^\circ C}} \times 100\%$$

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## 1.0 Functional Description

The LMK02000 precision clock conditioner combines the functions of jitter cleaning/reconditioning, multiplication, and distribution of a reference clock. The device integrates a high performance Integer-N Phase Locked Loop (PLL), three LVDS, and five LVPECL clock output distribution blocks.

Each clock distribution block includes a programmable divider, a phase synchronization circuit, a programmable delay, a clock output mux, and an LVDS or LVPECL output buffer. This allows multiple integer-related and phase-adjusted copies of the reference to be distributed to eight system components.

The clock conditioner comes in a 48-pin LLP package and is footprint compatible with other clocking devices in the same family.

### 1.1 BIAS PIN

To properly use the device, bypass Bias (pin 36) with a low leakage 1  $\mu$ F capacitor connected to Vcc. This is important for low noise performance.

### 1.2 LDO BYPASS

To properly use the device, bypass LDObyp1 (pin 9) with a 10  $\mu$ F capacitor and LDObyp2 (pin 10) with a 0.1  $\mu$ F capacitor.

### 1.3 OSCILLATOR INPUT PORT (OSCin, OSCin\*)

The purpose of OSCin is to provide the PLL with a reference signal. The OSCin port may be driven single ended by AC grounding OSCin\*.

### 1.4 FREQUENCY INPUT PORT (Fin, Fin\*)

The purpose of Fin is to provide the PLL with a feedback signal from an external oscillator. The Fin port may be driven single ended by AC grounding Fin\*.

### 1.5 CLKout DELAYS

Each individual clock output includes a delay adjustment. Clock output delay registers (CLKoutX\_DLY) support a 150 ps step size and range from 0 to 2250 ps of total delay.

### 1.6 LVDS/LVPECL OUTPUTS

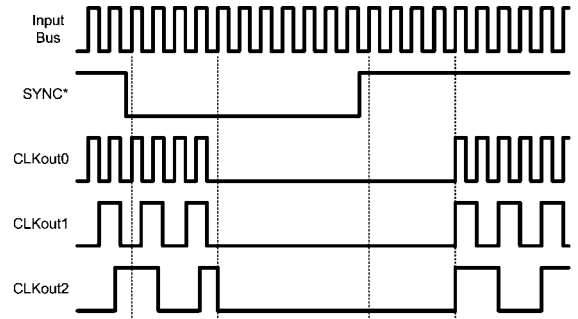
Each LVDS or LVPECL output may be disabled individually by programming the CLKoutX\_EN bits. All the outputs may be disabled simultaneously by pulling the GOE pin low or programming EN\_CLKout\_Global to 0.

## 1.7 GLOBAL CLOCK OUTPUT SYNCHRONIZATION

The SYNC\* synchronizes the clock outputs. When SYNC\* is held in a logic low state, the outputs are also held in a logic low state. When SYNC\* goes high, the clock outputs are activated and will transition to a high state simultaneously.

SYNC\* must be held low for greater than one clock cycle of the Input Channel Bus. Once this low event has been registered, the outputs will not reflect the low state for four more cycles. Similarly once SYNC\* becomes high, the outputs will not simultaneously transition high until four more Input Channel Bus clock cycles have passed. See the timing diagram below for further detail.

### SYNC\* Timing Diagram



20216504

## 1.8 GLOBAL OUTPUT ENABLE AND LOCK DETECT

Each clock output may be individually enabled. Each output enable control bit is gated with the Global Output Enable input pin (GOE) and the Global Output Enable bit (EN\_CLKout\_Global).

The GOE pin provides an internal pull-up. If it is unterminated externally, the clock output states are determined by the Clock Output Enable bits (CLKoutX\_EN).

All clock outputs can be disabled simultaneously if the GOE pin is pulled low by an external signal or EN\_CLKout\_Global is set to 0.

The Lock Detect (LD) signal can be connected to the GOE pin in which case all outputs are disabled automatically if the synthesizer is not locked.



## 2.0 General Programming Information

The LMK02000 device is programmed using sixteen 32-bit registers which control the device's operation. The registers consist of a data field and an address field. The last 4 register bits, ADDR[3:0] form the address field. The remaining 28 bits form the data field DATA[27:0].

During programming LE is low, serial data is clocked in on the rising edge of clock (MSB first). When LE goes high, data is transferred to the register bank selected by the address field. Only registers R0 to R7, R14, and R15 need to be programmed for proper device operation.

For the frequency calibration algorithm to work properly OSCin must be driven by a valid signal when R15 is programmed.

**2.1 LMK02000 REGISTER MAP**

Register	31	30	29	28	27	26	25	24	23	22	21	20	19	18	Data [27:0]										A3	A2	A1	A0									
																CLKout7_MUX [1:0]	CLKout6_MUX [1:0]	CLKout5_MUX [1:0]	CLKout4_MUX [1:0]	CLKout3_MUX [1:0]	CLKout2_MUX [1:0]	CLKout1_MUX [1:0]	CLKout0_MUX [1:0]	CLKout7_DIV [7:0]	CLKout6_DIV [7:0]	CLKout5_DIV [7:0]	CLKout4_DIV [7:0]	CLKout3_DIV [7:0]	CLKout2_DLY [3:0]	CLKout1_DLY [3:0]	CLKout0_DLY [3:0]						
R0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	
R1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	1
R2	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
R3	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	1
R4	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
R5	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	1
R6	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
R7	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	1

Register	31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
R14	0	0	0	0	EN_CLKout_Global	POWERDOWN	0	0		PLL_MUX [3:0]								PLL_R [11:0]						0	0	0	0	0	0	0	0	0
R15	PLL_CP_GAIN [1:0]	INPUT_DIV [3:0]													PLL_N [17:0]									0	0	0	0	0	0	0	0	0

## 2.2 REGISTER R0 to R7

Registers R0 through R7 control the eight clock output pins. Register R0 controls CLKout0, Register R1 controls CLKout1, and so on. Aside from this, the functions of these bits are identical. The X in CLKoutX\_MUX, CLKoutX\_DIV, CLKoutX\_DLY, and CLKoutX\_EN denote the actual clock output which may be from 0 to 7.

### 2.2.1 CLKoutX\_MUX[1:0] -- Clock Output Multiplexers

These bits control the Clock Output Multiplexer for each pin. Changing between the different modes changes the blocks in the signal path and therefore incurs a delay relative to the bypass mode. The different MUX modes and associated delays are listed below.

CLKoutX_MUX[1:0]	Mode	Added Delay Relative to Bypass Mode
0	Bypassed	0 ps
1	Divided	100 ps
2	Delayed	400 ps (In addition to the programmed delay)
3	Divided and Delayed	500 ps (In addition to the programmed delay)

### 2.2.2 CLKoutX\_DIV[7:0] -- Clock Output Dividers

These bits control the clock output divider value. In order for these dividers to be active, the respective CLKoutX\_MUX (See 2.2.1) bit must be set to either "Divided" or "Divided and Delayed" mode. After all the dividers are programmed, the SYNC\* pin must be used to ensure that all edges of the clock output pins are aligned (See 1.7). The Clock Output Dividers follow the Input Divider so the final clock divide for an output is Input Divider x Clock Output Divider. By adding the divider block to the output path a fixed delay of approximately 100 ps is incurred.

The actual Clock Output Divide value is twice the binary value programmed as listed in the table below.

CLKoutX_DIV[7:0]								Clock Output Divider value
0	0	0	0	0	0	0	0	Invalid
0	0	0	0	0	0	0	1	2
0	0	0	0	0	0	1	0	4
0	0	0	0	0	0	1	1	6
0	0	0	0	0	1	0	0	8
0	0	0	0	0	1	0	1	10
.	.	.	.	.	.	.	.	...
1	1	1	1	1	1	1	1	510

### 2.2.3 CLKoutX\_DLY[3:0] -- Clock Output Delays

These bits control the delay stages for each clock output pin. In order for these delays to be active, the respective CLKoutX\_MUX (See 2.2.1) bit must be set to either "Delayed" or "Divided and Delayed" mode. By adding the delay block to the output path a fixed delay of approximately 400 ps is incurred in addition to the delay shown in the table below.

CLKoutX_DLY[3:0]	Delay (ps)
0	0
1	150
2	300
3	450
4	600
5	750
6	900
7	1050
8	1200
9	1350
10	1500
11	1650
12	1800
13	1950
14	2100
15	2250

### 2.2.4 CLKoutX\_EN bit -- Clock Output Enables

This bit controls whether each clock output is enabled or not. If the EN\_CLKout\_Global bit (See 2.3.4) is set to zero or if GOE pin is held low, all CLKoutX\_EN bit states will be ignored and all clock outputs will be disabled.

CLKoutX_EN bit	EN_CLKout_Global bit	GOE pin	Clock X Output State
Don't care	Don't care	0	Disabled
Don't care	0	Don't care	Disabled
0	Don't care	Don't care	Disabled
1	1	High	Enabled

## 2.3 REGISTER R14

### 2.3.1 PLL\_R[11:0] -- R Divider Value

These bits program the PLL R Divider and are programmed in binary fashion.

PLL_R[11:0]											PLL R Divide Value	
0	0	0	0	0	0	0	0	0	0	0	0	Invalid
0	0	0	0	0	0	0	0	0	0	0	1	1
0	0	0	0	0	0	0	0	0	0	1	0	2
.	.	.	.	.	.	.	.	.	.	.	.	...
1	1	1	1	1	1	1	1	1	1	1	1	4095

### 2.3.2 PLL\_MUX[3:0] -- Multiplexer Control for LD Pin

These bits set the output mode of the LD pin. The table below lists several different modes.

PLL_MUX[3:0]	Output Type	LD Pin Function
0	Hi-Z	Disabled
1	Push-Pull	Logic High
2	Push-Pull	Logic Low
3	Push-Pull	Digital Lock Detect (Active High)
4	Push-Pull	Digital Lock Detect (Active Low)
5	Push-Pull	Analog Lock Detect
6	Open Drain NMOS	Analog Lock Detect
7	Open Drain PMOS	Analog Lock Detect
8	Push-Pull	N Divider Output (Very Low Duty Cycle)
9	Push-Pull	N Divider Output/2 (50% Duty Cycle)
10	Push-Pull	R Divider Output (Very Low Duty Cycle)
11	Push-Pull	R Divider Output/2 (50% Duty Cycle)
12 to 15	Invalid	

### 2.3.3 POWERDOWN bit -- Device Power Down

This bit can power down the device. Enabling this bit powers down the entire chip and all blocks, regardless of the state of any of the other bits or pins.

POWERDOWN bit	Mode
0	Normal Operation
1	Entire Chip Powered Down

### 2.3.4 EN\_CLKout\_Global bit -- Global Clock Output Enable

This bit overrides the individual CLKoutX\_EN bits (See 2.2.4). When this bit is set to 0, all clock outputs are disabled, regardless of the state of any of the other bits or pins.

EN_CLKout_Global bit	Clock Outputs
0	All Disabled
1	Normal Operation

### 2.3.5 EN\_Fout bit -- Fout port enable

This bit enables the Fout pin.

EN_Fout bit	Fout Pin Status
0	Disabled
1	Enabled

## 2.4 Register R15

### 2.4.1 PLL\_N[17:0] -- PLL N Divider

These bits program the divide value for the PLL N Divider. The PLL N Divider follows the Input Divider and precedes the PLL phase detector. Since the Input Divider is also in the feedback path from the VCO to the PLL Phase Detector, the total N divide value,  $N_{Total}$ , is also influenced by the Input Divider value.  $N_{Total} = \text{PLL N Divider} * \text{Input Divider}$ . The VCO frequency is calculated as,  $f_{VCO} = f_{OSCin} * \text{PLL N Divider} * \text{Input Divider} / R$ . Since the PLL N divider is a pure binary counter, there are no illegal divide values for PLL\_N[17:0].

PLL_N[17:0]																	PLL N Divider Value	
0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	Invalid
0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	1	1
.	.	.	.	.	.	.	.	.	.	.	.	.	.	.	.	.	.	...
1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	262143

### 2.4.2 INPUT\_DIV[3:0] -- Input Divider

These bits program the divide value for the Input Divider. The Input Divider follows the VCO output and precedes the clock distribution blocks. Since the Input Divider is in the feedback path from the VCO to the PLL phase detector the Input Divider contributes to the total N divide value,  $N_{Total}$ .  $N_{Total} = \text{PLL N Divider} * \text{Input Divider}$ . The Input Divider can not be bypassed. See 2.4.1 (PLL N Divider) for more information on setting the VCO frequency.

INPUT_DIV[3:0]				Input Divider Value
0	0	0	0	Invalid
0	0	0	1	Invalid
0	0	1	0	2
0	0	1	1	3
0	1	0	0	4
0	1	0	1	5
0	1	1	0	6
0	1	1	1	7
1	0	0	0	8
1	0	0	1	Invalid
.	.	.	.	...
1	1	1	1	Invalid

### 2.4.3 PLL\_CP\_GAIN[1:0] -- PLL Charge Pump Gain

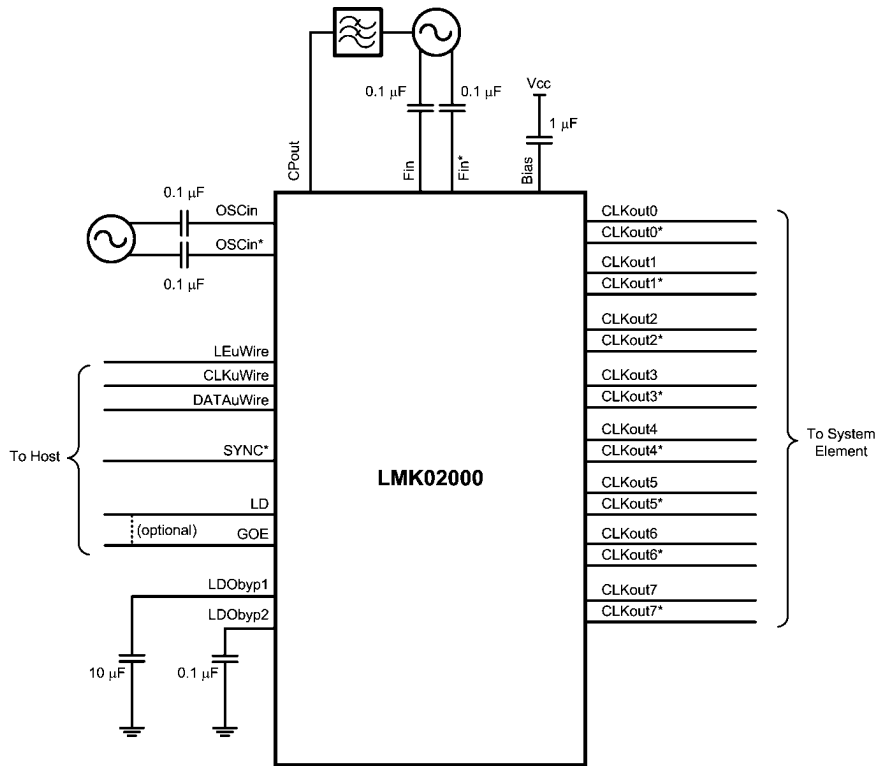
These bits set the charge pump gain of the PLL.

PLL_CP_GAIN[1:0]	Charge Pump Gain
0	1x
1	4x
2	16x
3	32x

### 3.0 Application Information

#### 3.1 System Level Diagram

The following shows the LMK02000 in a typical application. In this setup the clock may be multiplied, reconditioned, and redistributed.



20216570

#### 3.2 Bias Pin

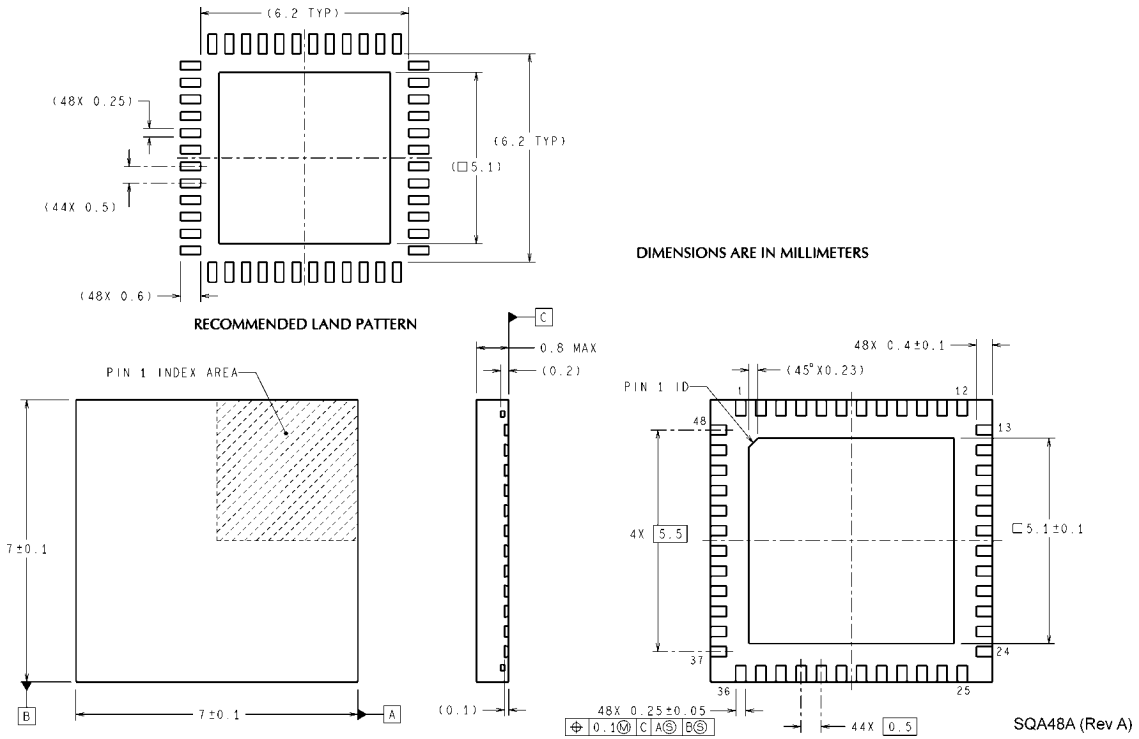
To properly use the device, bypass Bias (pin 36) with a low leakage 1 μF capacitor connected to Vcc. This is important for low noise performance.

#### 3.3 LDO bypass

To properly use the device, bypass LDObyp1 (pin 9) with a 10 μF capacitor and LDObyp2 (pin 10) with a 0.1 μF capacitor.



**Physical Dimensions** inches (millimeters) unless otherwise noted



**Leadless Leadframe Package (Bottom View)  
48 Pin LLP (SQA48A) Package**

Order Number	Package Marking	Packing	LVDS Outputs	LVPECL Outputs
LMK02000	K02000 I	1000 Unit Tape and Reel	3	5
LMK02000X	K02000 I	4000 Unit Tape and Reel	3	5

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