

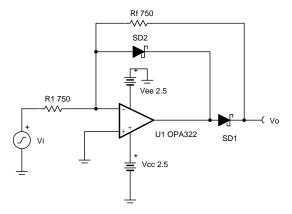
# Half-wave rectifier circuit

#### **Design Goals**

Input		Output		Supply	
$V_{iMin}$	$V_{iMax}$	$V_{oMin}$	V <sub>oMax</sub>	V <sub>cc</sub>	V <sub>ee</sub>
±0.2mV <sub>pp</sub>	±4V <sub>pp</sub>	0.1V <sub>p</sub>	2V <sub>p</sub>	2.5V	-2.5V

#### **Design Description**

The precision half-wave rectifier inverts and transfers only the negative-half input of a time varying input signal (preferably sinusoidal) to its output. By appropriately selecting the feedback resistor values, different gains can be achieved. Precision half-wave rectifiers are commonly used with other op amp circuits such as a peak-detector or bandwidth limited non-inverting amplifier to produce a DC output voltage. This configuration has been designed to work for sinusoidal input signals between  $0.2 \text{mV}_{pp}$  and  $4 \text{V}_{pp}$  at frequencies up to 50 kHz.



## **Design Notes**

- 1. Select an op amp with a high slew rate. When the input signal changes polarities, the amplifier output must slew two diode voltage drops.
- 2. Set output range based on linear output swing (see A<sub>ol</sub> specification).
- Use fast switching diodes. High-frequency input signals will be distorted depending on the speed by
  which the diodes can transition from blocking to forward conducting mode. Schottky diodes might be a
  preferable choice, since these have faster transitions than pn-junction diodes at the expense of higher
  reverse leakage.
- 4. The resistor tolerance sets the circuit gain error.
- 5. Minimize noise errors by selecting low-value resistors.



## **Design Steps**

1. Set the desired gain of the half-wave rectifier to select the feedback resistors.

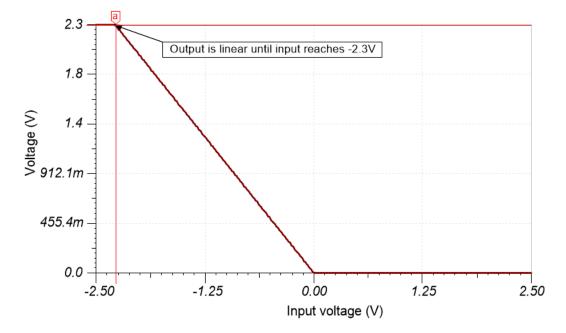
$$\begin{split} &V_o = Gain \times V_i \\ &Gain = -\frac{R_f}{R_1} = -1 \\ &R_f = R_1 = 2 \times R_{eq} \end{split}$$

- Where  $R_{eq}$  is the parallel combination of  $R_1$  and  $R_f$
- 2. Select the resistors such that the resistor noise is negligible compared to the voltage broadband noise of the op amp.

$$\begin{split} &E_{nr} = \sqrt{4 \times k_b \times T \times R_{eq}} \\ &R_{eq} \leq \frac{E_{nbb}^2}{4 \times k_b \times T \times 3^2} = (Enbb) \\ &= 7 \cdot 5 \frac{nV}{\sqrt{Hz}} = \frac{(7.5 \times 10^{-9})^2}{4 \times 1.381 \times 10^{-23} \times 298 \times 3^2} = 380\Omega \\ &R_f = R_1 \leq 760\Omega \rightarrow 750\Omega \text{ (Standard Value)} \end{split}$$

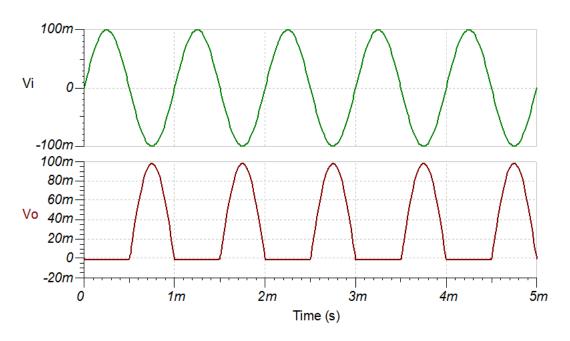
### **Design Simulations**

#### **DC Simulation Results**

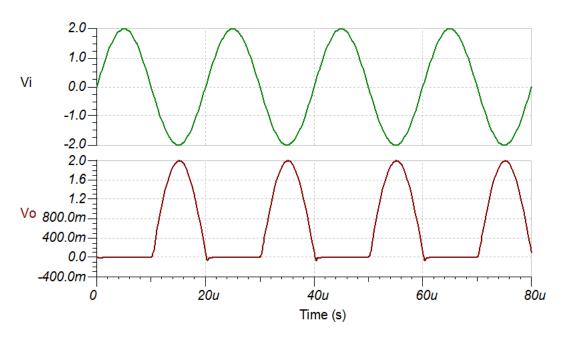




## **Transient Simulation Results**



 $200 \mathrm{mV}_\mathrm{pp}$  at 1kHz



 $\mathrm{2V}_{\mathrm{pp}}$  at  $\mathrm{50kHz}$ 



## **Design References**

See *Analog Engineer's Circuit Cookbooks* for TI's comprehensive circuit library. See circuit SPICE simulation file SBOC509.

## **Design Featured Op Amp**

OPA322				
V <sub>ss</sub>	1.8V to 5.5V			
V <sub>inCM</sub>	Rail-to-rail			
V <sub>out</sub>	Rail-to-rail			
V <sub>os</sub>	500μV			
I <sub>q</sub>	1.6mA/Ch			
I <sub>b</sub>	0.2pA			
UGBW	20MHz			
SR	10V/µs			
#Channels	1, 2, 4			
www.ti.com/product/opa322				

## **Design Alternate Op Amp**

OPA2325				
$V_{ss}$	2.2V to 5.5V			
V <sub>inCM</sub>	Rail-to-rail			
V <sub>out</sub>	Rail-to-rail			
V <sub>os</sub>	40μV			
I <sub>q</sub>	0.65mA/Ch			
I <sub>b</sub>	0.2pA			
UGBW	10MHz			
SR	5V/μs			
#Channels	2µ			
www.ti.com/product/opa2325				

## **Revision History**

Revision	Date	Change
Α	January 2019	Downscale the title and changed title role to 'Amplifiers'.
		Added link to circuit cookbook landing page and link to Spice simulation file.

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